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Fig. 1 shows the Cadence schematic used for this problem. The I and Q paths were included in the schematic so a realistic load was presented to the output of the RF mixer.

- Using a PSS analysis and sweeping  $L_1$  and  $L_2$ , we get  $L_1 = 4.91 \text{ nH}$  and  $L_2 = 23.59 \text{ nH}$  as the inductance values necessary for resonance at their respective frequencies. Fig. 2 shows the spectrum at the output of the LNA with resonance at 5.2 GHz (when driven by an input power of  $-40 \text{ dBm}$ ). Fig. 3 shows the spectrum at the output of the RF mixer with resonance at 1.7 GHz.
- Fig. 4 shows the spectrum at the baseband output,  $V_{BB}$ , when the input is driven by a  $-40 \text{ dBm}$  (or  $6.325 \text{ mV}$  amplitude) signal. From this, we can see that the amplitude of the baseband signal is  $12.26 \text{ mV}$ . The total voltage gain is then:

$$A_v = \frac{12.26 \text{ mV}}{\frac{1}{2}(6.325 \text{ mV})} = 3.877 = \boxed{11.77 \text{ dB}} \quad (9)$$

- Using an input signal at  $1.8 \text{ GHz}$  (the image frequency) with a power of  $-40 \text{ dBm}$ , the output spectrum is as shown in Fig. 5. The amplitude of the baseband output due to the image is  $2.047 \text{ mV}$ .

Since the signal and the image were input at the same power, the image reject ratio is simply the image-to-signal ratio at the output. The signal amplitude, from part (b), is  $12.26 \text{ mV}$ , giving:

$$IRR = \frac{2.047 \text{ mV}}{12.26 \text{ mV}} = 16.70 = \boxed{-15.55 \text{ dB}} \quad (5)$$

- These are two mechanisms by which a  $8.7 \text{ GHz}$  interferer can be translated to baseband:

- The interferer mixes with the first LO, giving a component at  $\omega_{int} - \omega_{LO1} = 8.7 \text{ GHz} - 3.5 \text{ GHz} = 5.2 \text{ GHz}$ , which appears at the output of the RF mixer. This component mixes with the 3rd harmonic of the second LO, giving a component at  $3\omega_{LO2} - 5.2 \text{ GHz} = 5.25 \text{ GHz} - 5.2 \text{ GHz} = 50 \text{ MHz}$ , which is the baseband frequency.
- The interferer mixes with the 3rd harmonic of the first LO, giving a component at  $3\omega_{LO1} - \omega_{int} = 10.5 \text{ GHz} - 8.7 \text{ GHz} = 1.8 \text{ GHz}$ , which appears at the output of the RF mixer. This component mixes with the second LO, giving a component at  $1.8 \text{ GHz} - \omega_{LO2} = 1.8 \text{ GHz} - 1.75 \text{ GHz} = 50 \text{ MHz}$ , which is the baseband frequency.

Running a simulation with an input at  $8.7 \text{ GHz}$  with an input power of  $-40 \text{ dBm}$  gives the baseband spectrum shown in Fig. 6. From this spectrum we can see that the output amplitude is  $539.1 \mu\text{V}$ , giving a gain of:

$$A_{v,int} = \frac{539.1 \mu\text{V}}{6.325 \text{ mV}} = 0.0852 = \boxed{-21.39 \text{ dB}}$$

Normalized to the gain from (b), we have:

$$\frac{A_{v,int}}{A_{v,sig}} = \frac{0.0852}{3.877} = 0.0220 = \boxed{-33.16 \text{ dB}} \quad (10)$$

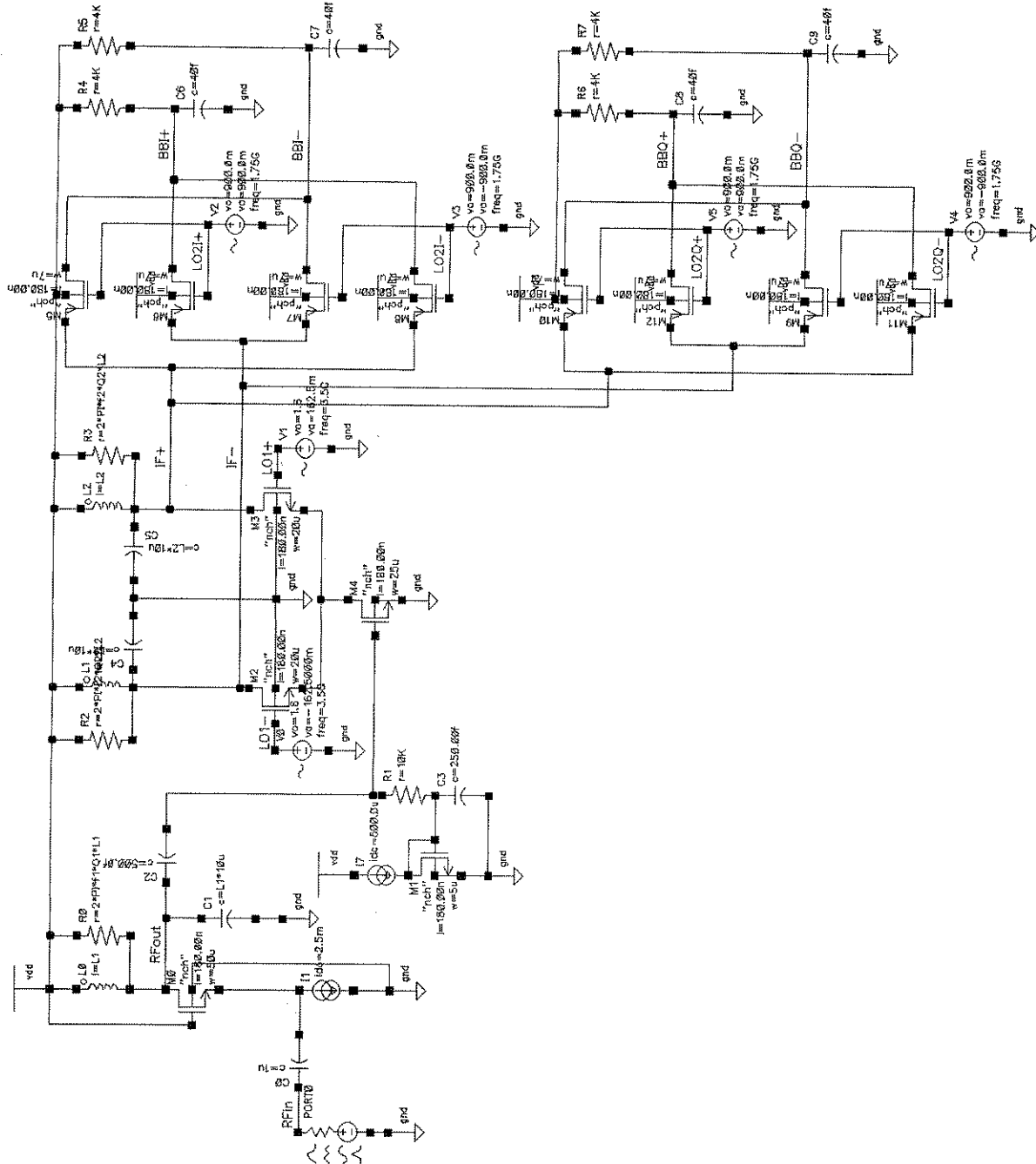


Figure 1: Cadence schematic

Periodic Steady State Response

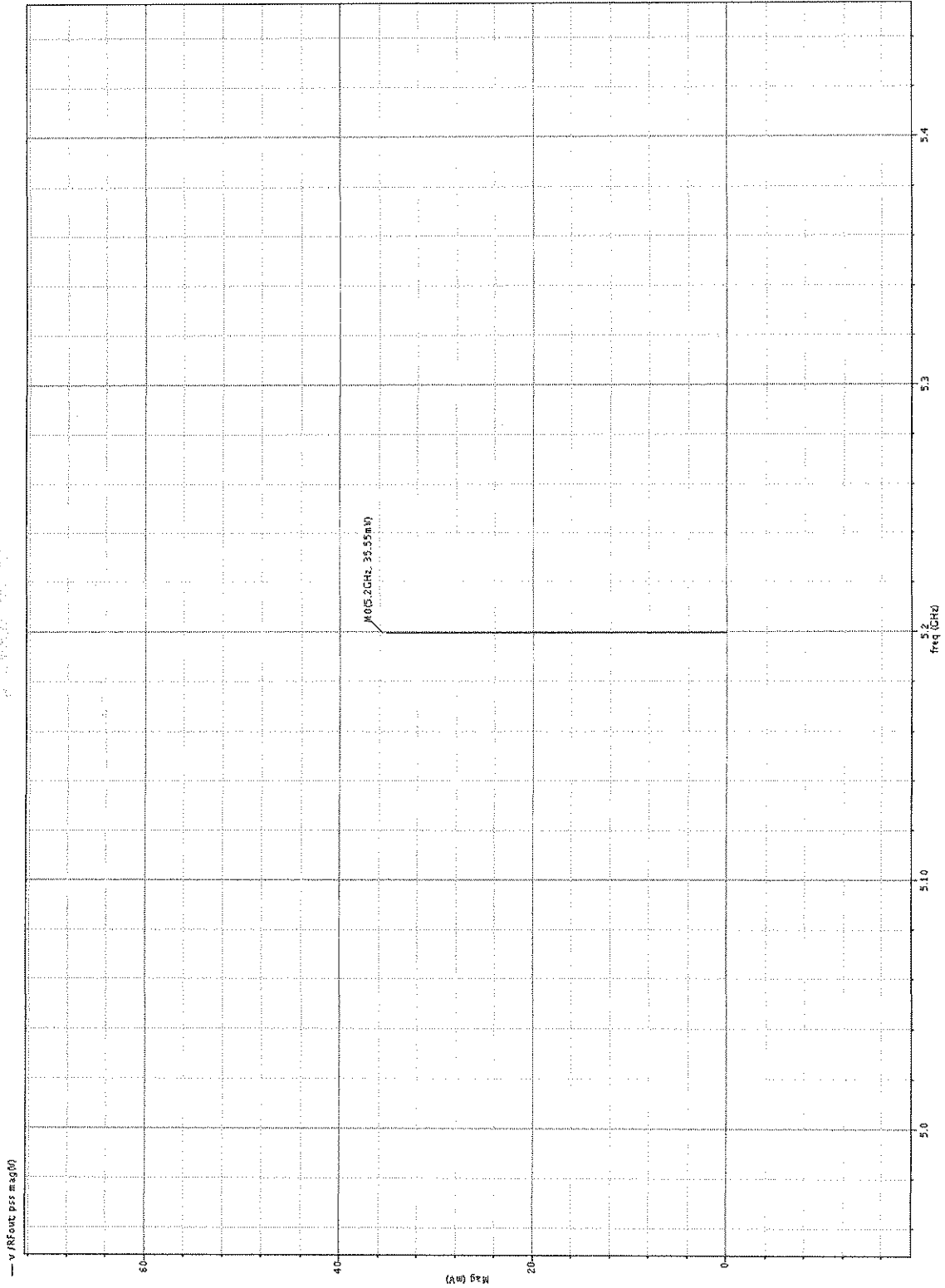


Figure 2: Spectrum at the output of the LNA showing resonance at  $\omega_{RF} = 5.2$  GHz

Periodic Steady State Response

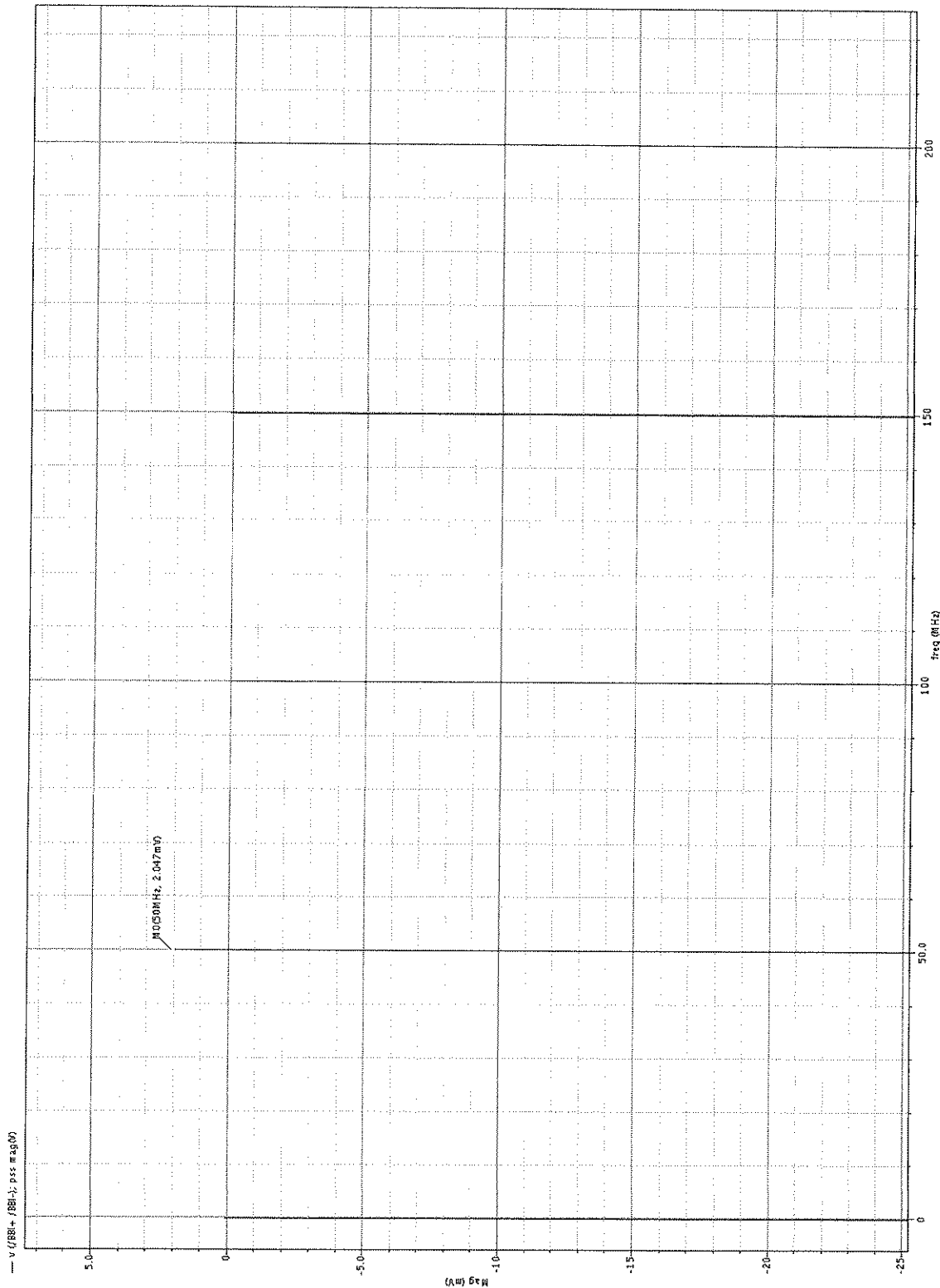


Figure 5: Baseband spectrum due to an input at the image frequency  $\omega_{im} = 1.8$  GHz with an input power of  $-40$  dBm

Periodic Steady State Response

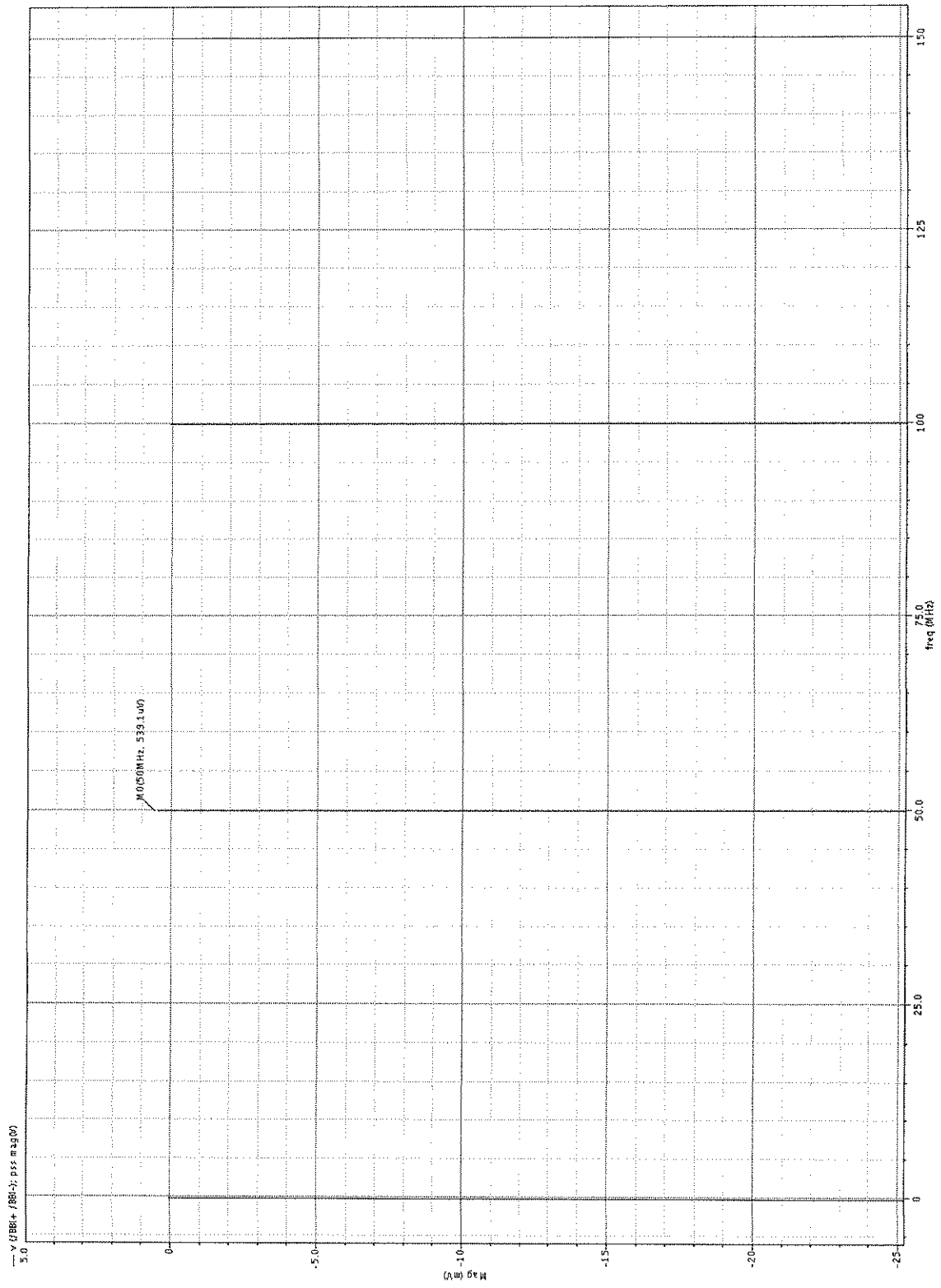


Figure 6: Baseband spectrum due to an interferer at  $\omega_{int} = 8.7$  GHz with an input power of  $-40$  dBm

